

# **50 Gbits/sec: The Next Mainstream Wireline Interconnect Lane Bit Rate**

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Forum 4: Emerging Short-Reach and  
High-Density Interconnect Solutions for  
Internet of Everything

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**FINISAR®**

# Outline

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- **Wireline Overview**
- 10 Gb/s Lane Optical Interfaces
- 25 Gb/s Lane Optical Interfaces
- 50 Gb/s Modulation Selection
- 50 Gb/s Lane Optical Interfaces
- Laser Driver IC
- Trans-Impedance Amplifier IC
- Clock Data Recovery IC
- ADC/DSP IC
- Summary
- References

# Major Applications

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- Ethernet (Datacom)
  - IEEE 802.3 standards
  - Mainstream High-volume Interfaces
- Transport Clients (Telecom)
  - ITU-T standards
  - Variants of Ethernet interfaces
- FibreChannel
  - T11 standards
  - Storage
- InfiniBand
  - Low latency is key requirement (ex. no FEC)
  - HPC (High Performance Computing)

# Ethernet Data Rates ( $\geq 10$ Gb/s)

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- Ethernet data rate is set by the rate of the MAC (Media Access Controller)
- Existing Ethernet data rate progression (Gb/s):  
 $10 \rightarrow 40 \rightarrow 100$
- Rates in standardization by IEEE 802.3:
  - 25 Gb/s (nearly completed)
  - 50 Gb/s (just started)
  - 200 Gb/s (just started)
  - 400 Gb/s
- Resulting Ethernet data rate progression (Gb/s):  
 $10 \rightarrow 25 \rightarrow 50 \text{ (& } 40\text{)} \rightarrow 100 \rightarrow 200 \rightarrow 400$

# Lane Rates & Technology ( $\geq 10$ Gb/s)

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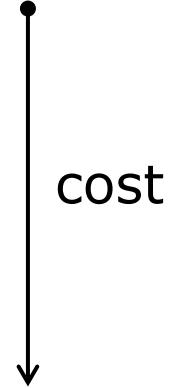
- In volume use
  - 10 Gb/s: 10Gbaud NRZ w/o FEC
  - 25 Gb/s: 25Gbaud NRZ w/o & w/ FEC
- In development for near-term volume use
  - **50 Gb/s: 25Gbaud PAM4 w/ FEC<sup>1</sup>**
- In development for near-term specialty apps.
  - 50 Gb/s: 50GBaud NRZ w/o FEC<sup>2</sup>
- In development for long-term use
  - 100 Gb/s: 50GBaud PAM4 w/ FEC<sup>2</sup>
  - 100 Gb/s: Complex Mod., ex. DMT, w/ FEC<sup>2</sup>

<sup>1</sup> Presentation focus

<sup>2</sup> Not discussed in this presentation

# Ethernet Optics Designations & Reach

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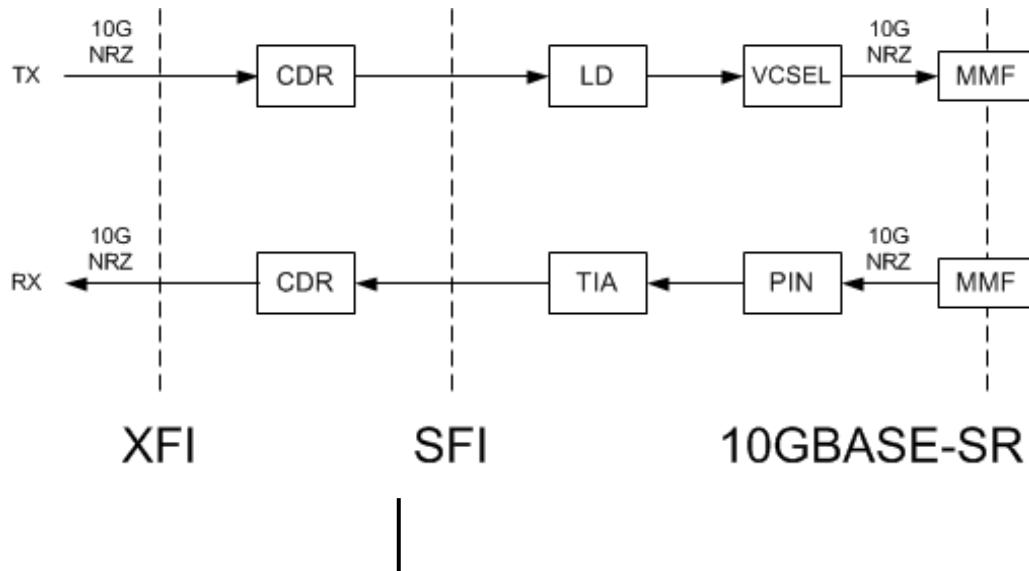
- SRn:  $\leq$  100m to 300m MMF (& SWDMn)
  - DRn:  $\leq$  500m SMF (& PSMn)
  - FRn:  $\leq$  2km SMF (& CWDMn)
  - LRn:  $\leq$  10km SMF
  - ERn:  $\leq$  40km SMF
  - “n” designates number of lanes, either parallel fiber pairs or duplex wavelengths
  - MMF: Multi-Mode Fiber, for lowest cost lasers
  - SMF: Single-Mode Fiber, for longer reaches
  - LC: Lucent Connector, duplex (2x) connector
  - MPO: Multi-fiber Push-On, parallel connector
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# 10GBASE-SR 850nm MMF Optics



Optics ICs:

- LD: Laser Driver
- TIA: Trans-Impedance Amp
- CDR: Clock Data Recovery

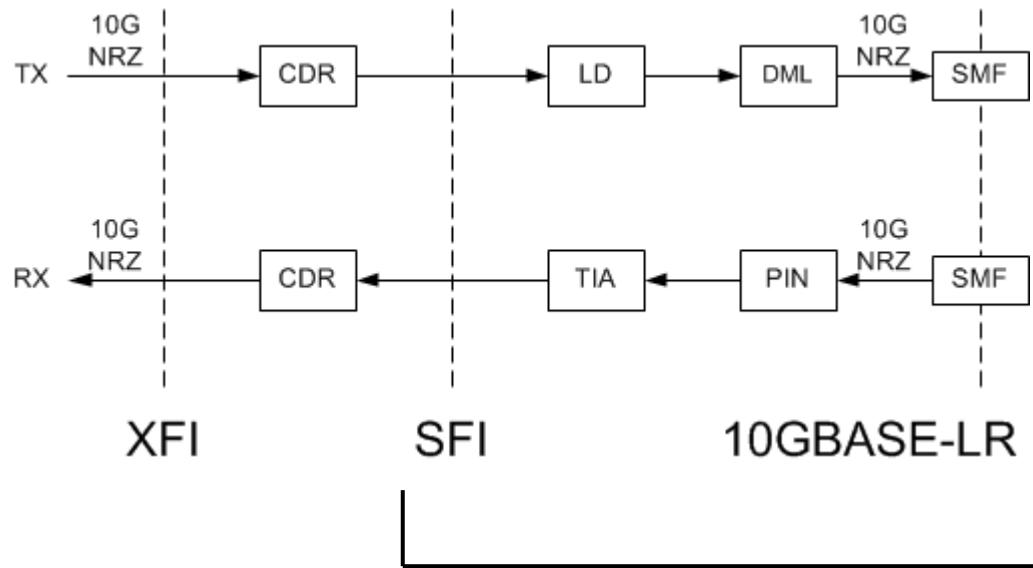
(optics defined on next page)

Dominant,  
standard form  
factor: SFP+  
w/ 2x LC



Lane Rate	No. of Lanes	Data Rate	
Gb/s	fiber pairs	$\lambda$	Gb/s
10	1	1	10

# 10GBASE-LR 1310nm SMF Optics



## Optics:

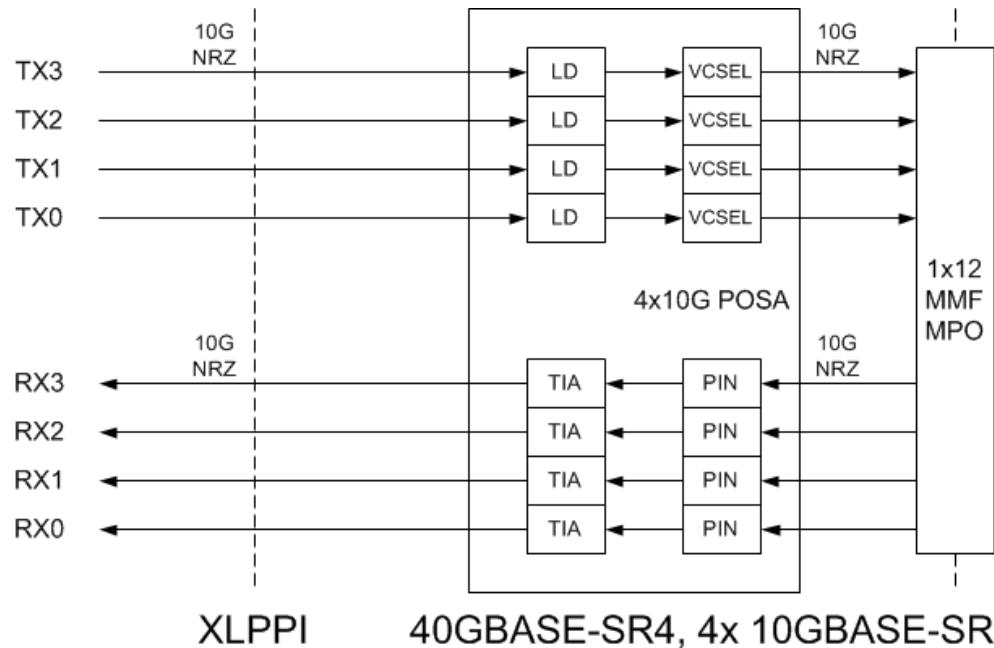
- VCSEL: Vertical Cavity Surface Emitting Laser
- DML: Directly Modulated Laser
- PIN: p-type | intrinsic | n-type (photodiode)

Dominant,  
standard form  
factor: SFP+  
w/ 2x LC



Lane Rate	No. of Lanes	Data Rate	
Gb/s	fiber pairs	$\lambda$	Gb/s
10	1	1	10

# 40GBASE-SR4 850nm MMF Optics



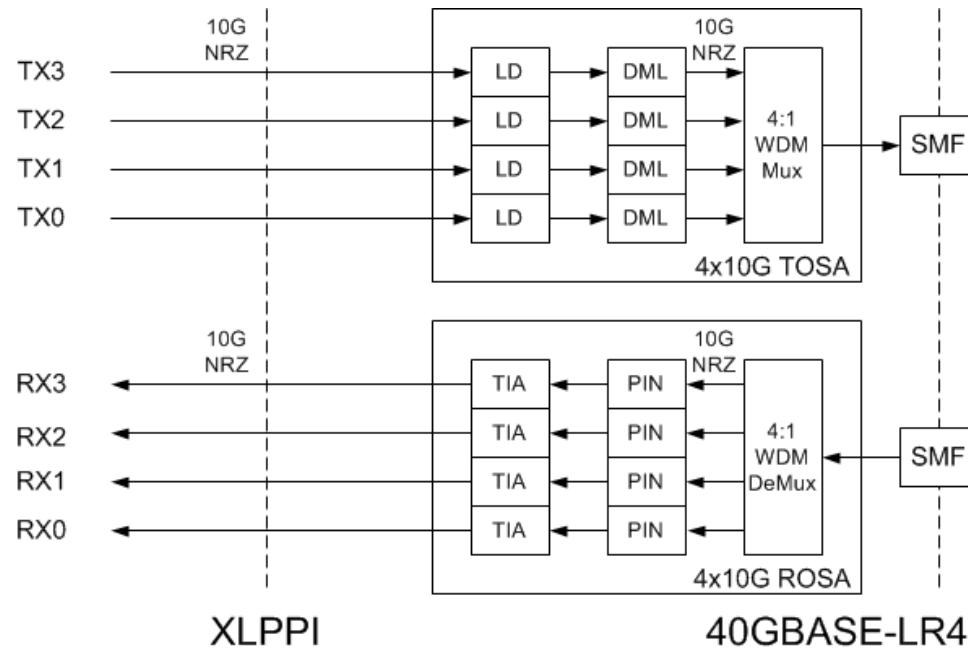
40G PSM4 is  
Parallel SMF  
version

Dominant,  
standard form  
factor: QSFP+  
w/ MPO



Lane Rate	No. of Lanes	Data Rate	
Gb/s	fiber pairs	$\lambda$	Gb/s
10	4	1	40

# 40GBASE-LR4 1310nm SMF Optics



40G SWDM4  
is WDM MMF  
version

Dominant,  
standard form  
factor: QSFP+  
w/ 2x LC



Lane Rate		No. of Lanes		Data Rate	
Gb/s	fiber pairs	$\lambda$	Gb/s		
10	1	4	40		

# 10 Gb/s Lane Optical Interfaces

Lane Rate	No. of Lanes	Data Rate	SW code	LW code	
Gb/s	fiber pairs	$\lambda$	Gb/s	(MMF)	(SMF)
10	1	1	10	<b>SR</b>	<b>LR</b>
10	4	1	40	<b>SR4</b>	<i>PSM4</i>
10	1	4	40	<i>SWDM4</i>	<b>LR4</b>
10	10	1	100	<b>SR10</b>	

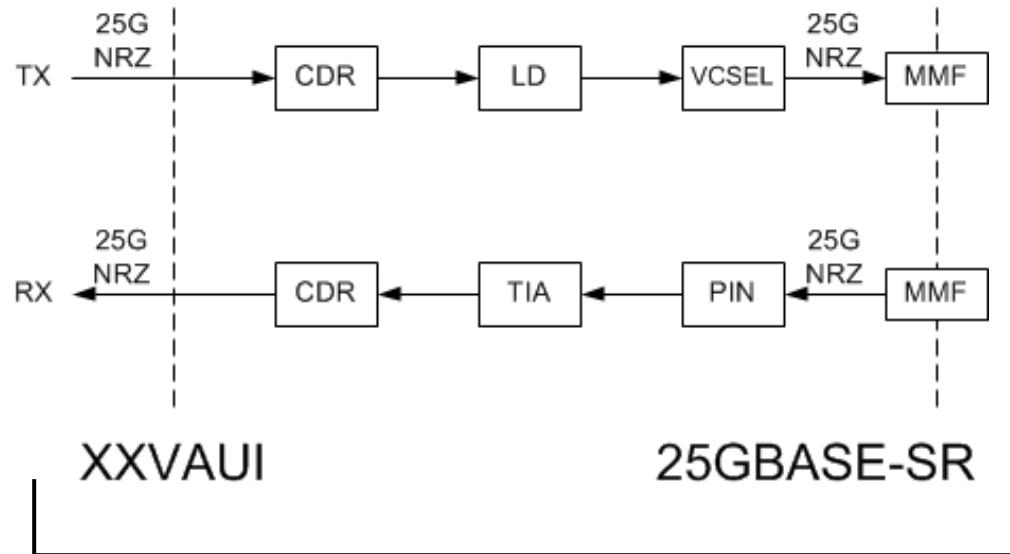
IEEE standards in BOLD; MSA or proprietary in *ITALICS*

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- **25 Gb/s Lane Optical Interfaces**
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# 25GBASE-SR 850nm MMF Optics

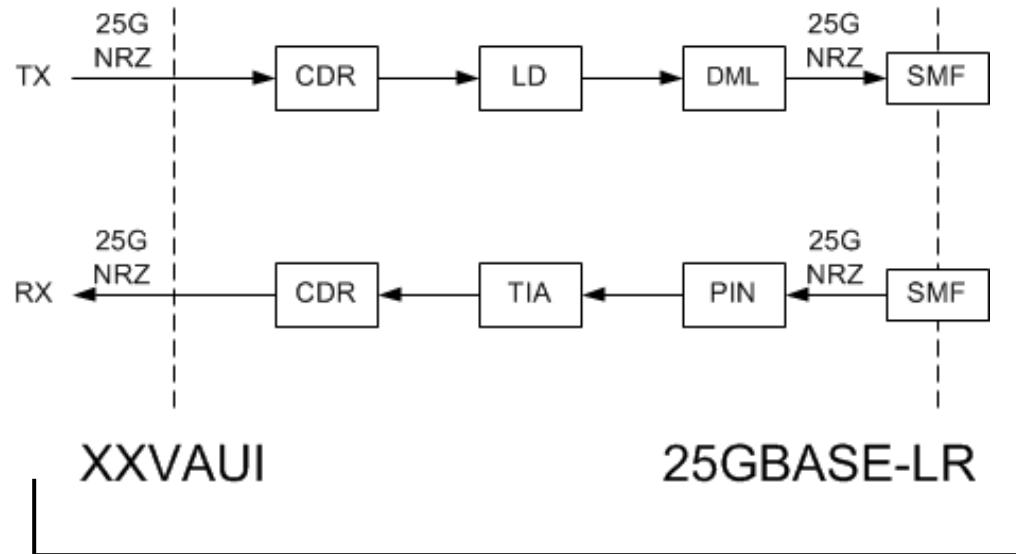


Dominant,  
standard form  
factor: SFP28  
w/ 2x LC



Lane Rate	No. of Lanes	Data Rate	
Gb/s	fiber pairs	$\lambda$	Gb/s
25	1	1	25

# 25GBASE-LR 1310nm SMF Optics

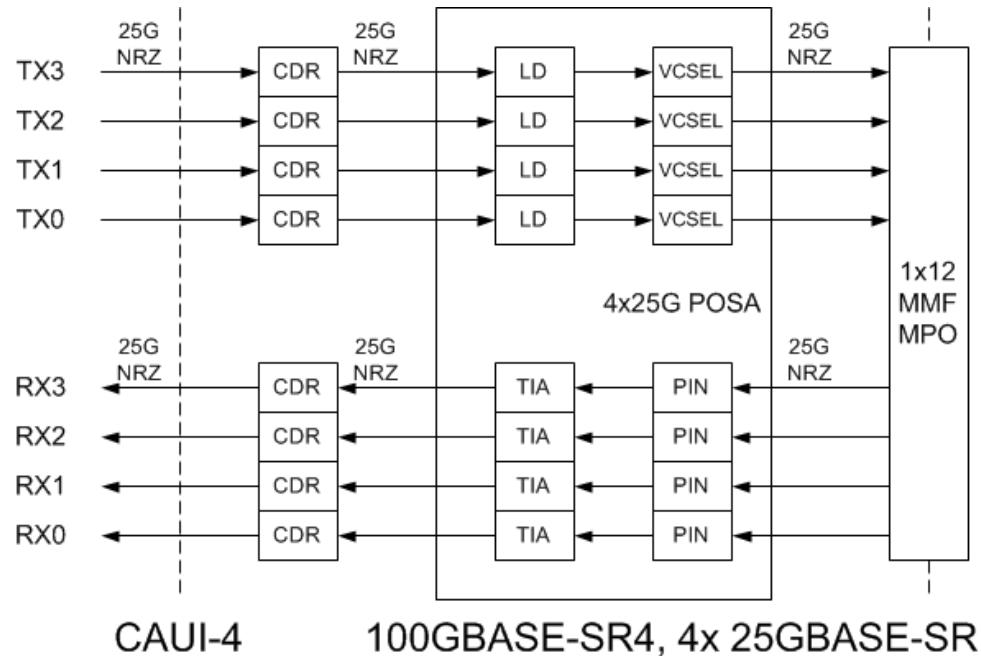


Dominant,  
standard form  
factor: SFP28  
w/ 2x LC



Lane Rate	No. of Lanes	Data Rate	
Gb/s	fiber pairs	$\lambda$	Gb/s
25	1	1	25

# 100GBASE-SR4 850nm MMF Optics



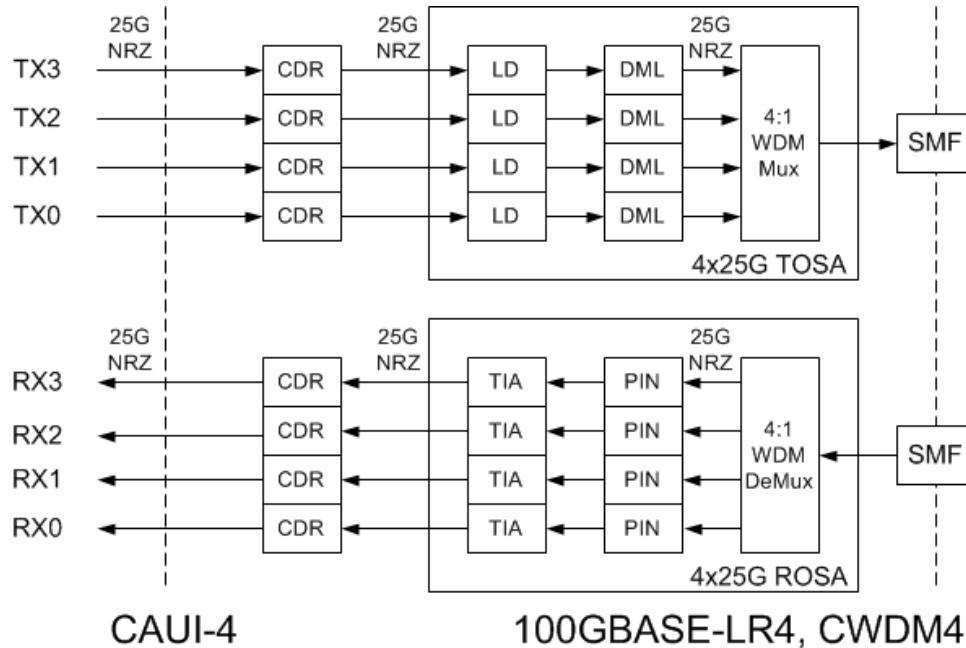
Dominant,  
standard form  
factor: QSFP28  
w/ MPO



100G PSM4 is  
parallel SMF  
version

Lane Rate	No. of Lanes	Data Rate	
Gb/s	fiber pairs	$\lambda$	Gb/s
25	4	1	100

# 100GBASE-LR4 1310nm SMF Optics



Dominant,  
standard form  
factor: QSFP28  
w/ 2x LC



100G SWDM4  
is WDM MMF  
version

Lane Rate	No. of Lanes	Data Rate	
Gb/s	fiber pairs	$\lambda$	Gb/s
25	1	4	100

# 25 Gb/s Lane Optical Interfaces

Lane Rate	No. of Lanes	Data Rate	SW code	LW code	
Gb/s	fiber pairs	$\lambda$	Gb/s	(MMF)	(SMF)
25	1	1	25	<b>SR</b>	<b>LR</b>
25	4	1	100	<b>SR4</b>	<i>PSM4</i>
25	1	4	100	<i>SWDM4</i>	<b>LR4</b>
25	8	1	200	<b>SR8</b>	<i>PSM8</i>
25	16	1	400	<b>SR16</b>	

IEEE standards in BOLD; MSA or proprietary in *ITALICS*

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# Shannon-Hartley Theorem

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$$C = B \log_2 (1 + S/N)$$

$C \triangleq$  Channel capacity

$B \triangleq$  Bandwidth

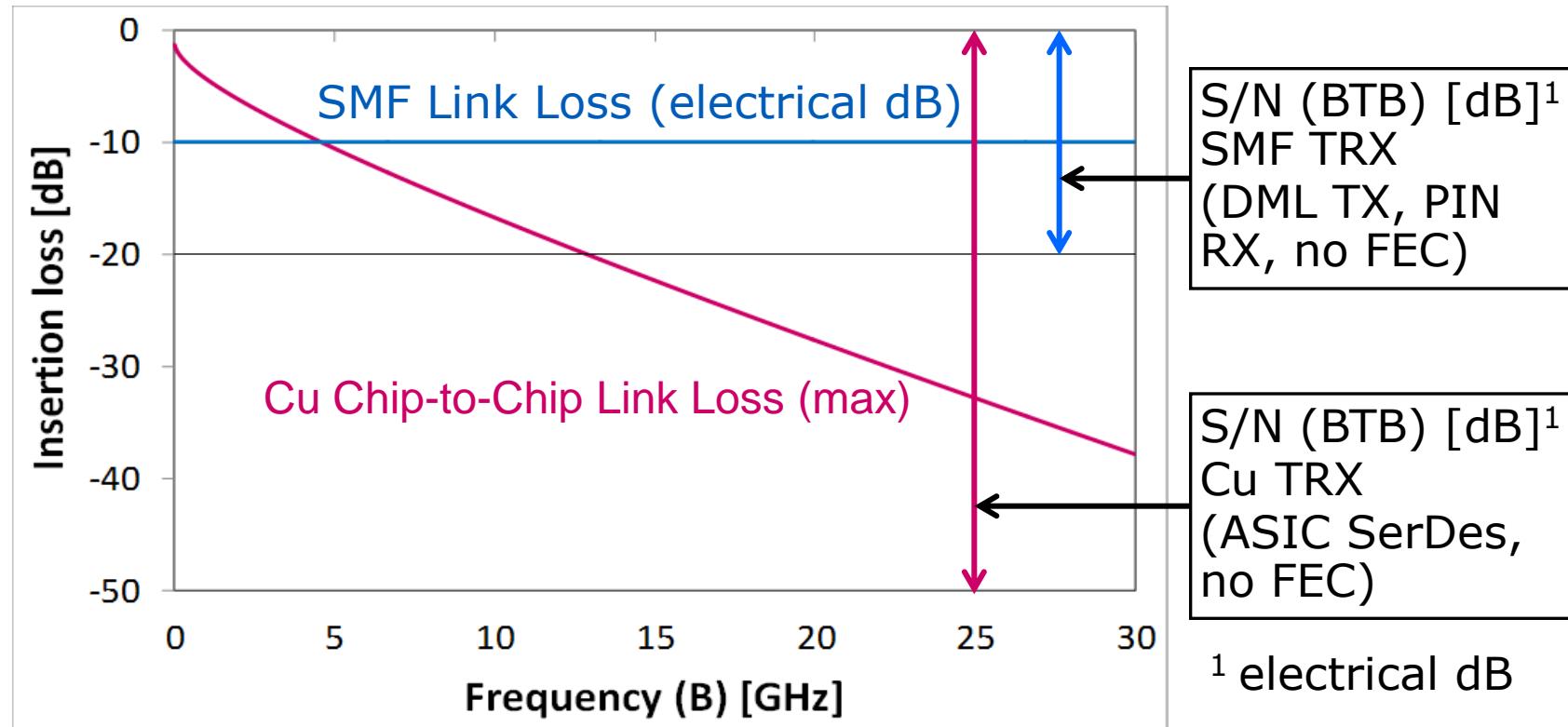
$S \triangleq$  Signal Power

$N \triangleq$  Noise Power

Guidance to increase C:

- If B limited, increase S/N to increase modulation order, i.e. more bits/Baud
- If S/N limited, increase B to increase Baud rate, i.e. switch faster

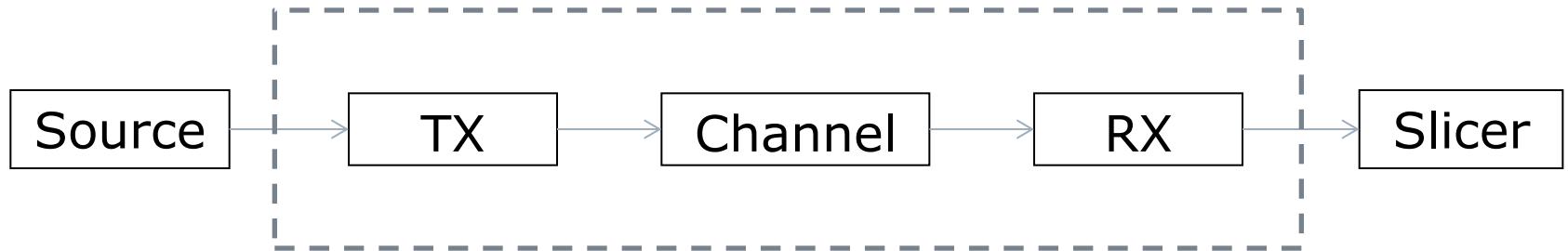
# Cu vs. SMF Link Loss & TRX S/N



- Cu channel limitation: Bandwidth (B)
- SMF channel limitation: S/N

# Ideal SMF Link Model

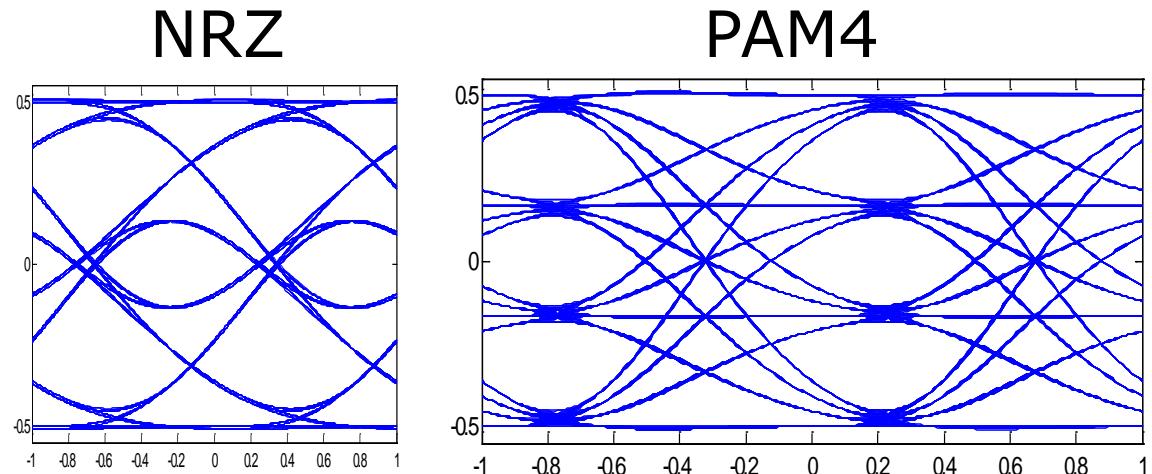
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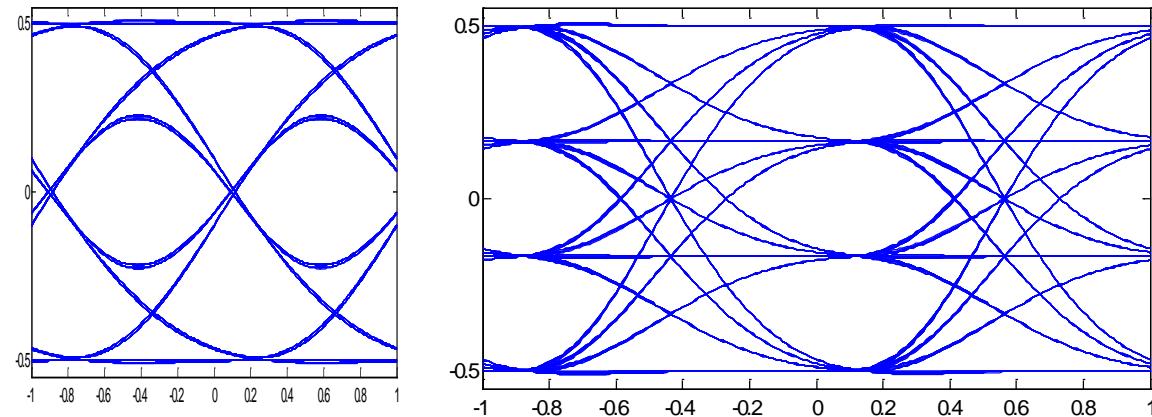
- SMF channel assumed ideal
- 4<sup>th</sup> order BT filter model for TX \* Channel \* RX
- Bandwidth  $\triangleq B = \alpha * \text{bit-rate}$
- Example bandwidths for bit rate = 56Gb/s
  - ex. 1:  $\alpha = 0.25 \rightarrow B = 14\text{GHz}$
  - ex. 2:  $\alpha = 0.30 \rightarrow B = 17\text{GHz}$

# Slicer Input of Ideal SMF Link

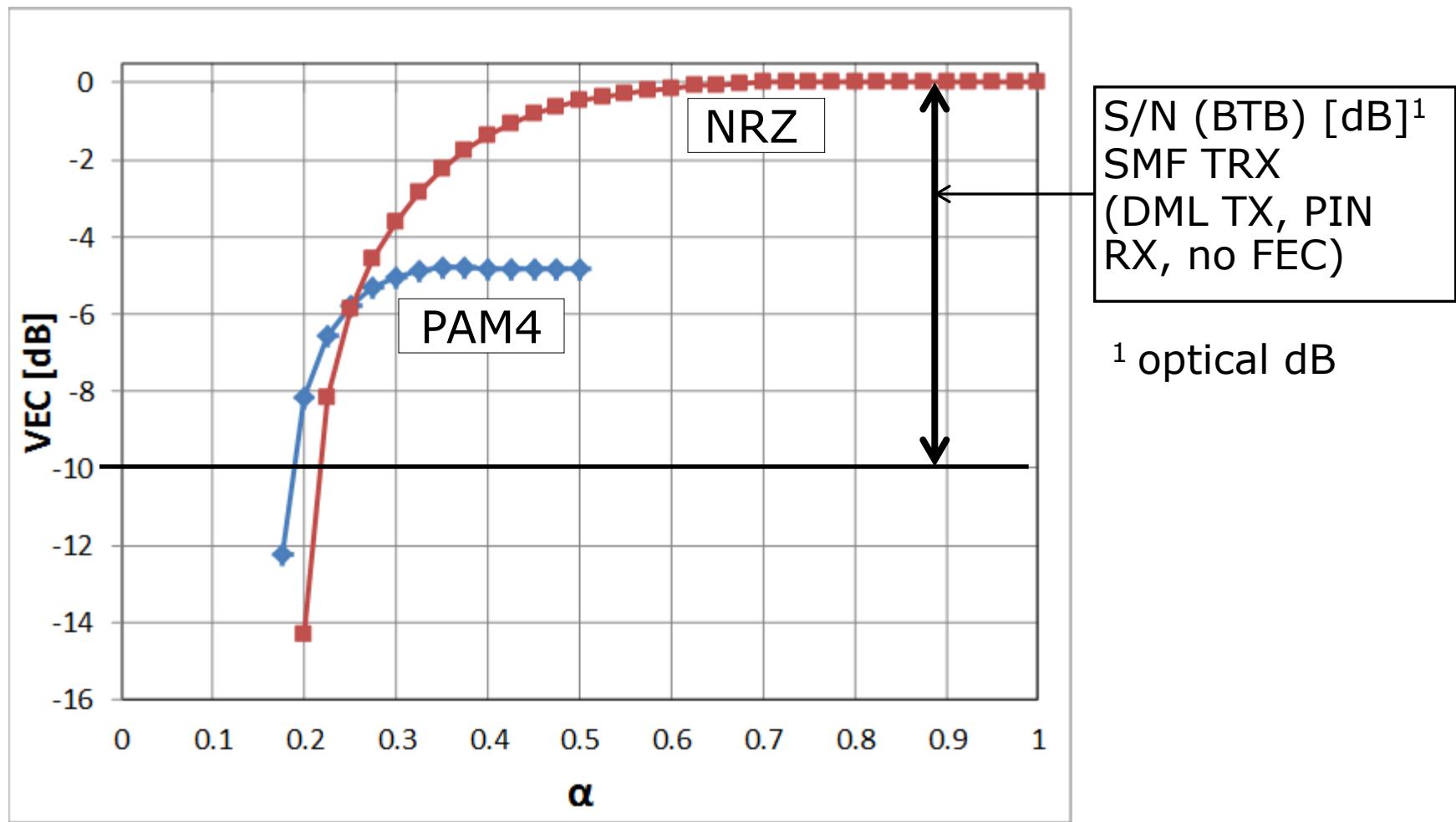
ex. 1  
 $a = 0.25$   
(14 GHz)



ex. 2  
 $a = 0.30$   
(17 GHz)



# Vertical Eye Closure at Slicer Input



# 50 Gb/s NRZ vs. PAM4 Optical Lanes

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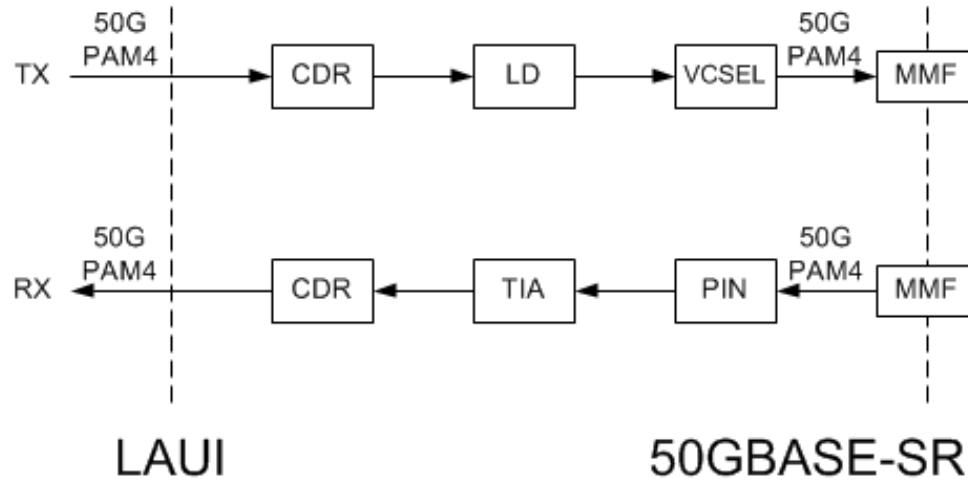
- 50G NRZ Advantages:
  - Optical SNR
  - Well understood development methodology  
ex. 10G NRZ → 25G NRZ
- 50G PAM4 Advantages:
  - 50G PAM4 IC ecosystem & volume
  - 25G NRZ optical packaging reuse
  - 25G NRZ SMF & MMF laser reuse
- Deciding factor in favor of 50G PAM4
  - Optics is the tail on the IC industry dog
  - PAM4 was developed for 50G ASIC SerDes because of channel bandwidth limitations
  - Despite SNR limitations optics tail wagged

# Outline

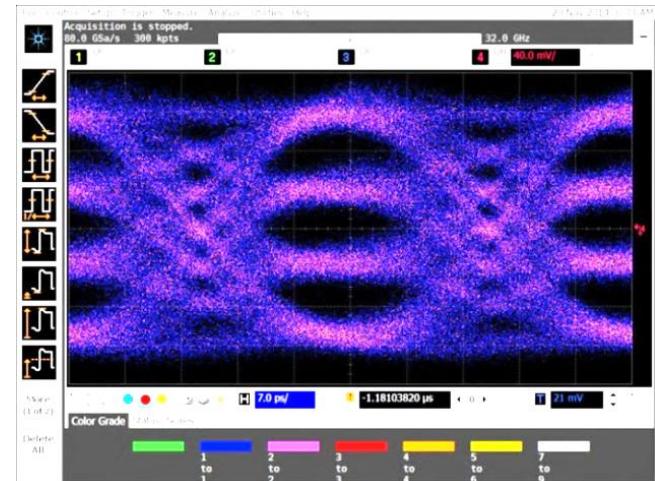
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# 50GBASE-SR 850nm MMF Optics



56Gb/s PAM4 optical eye

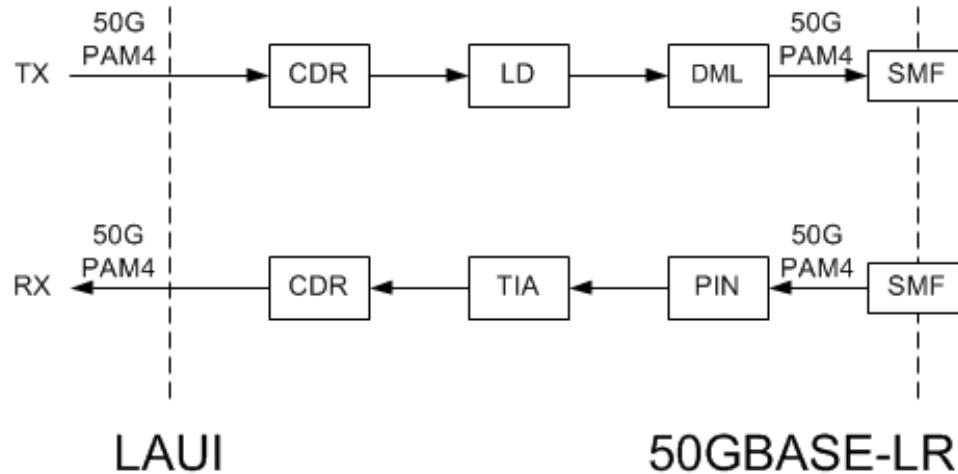


Likely dominant,  
standard form  
factor: SFP28  
w/ 2x LC

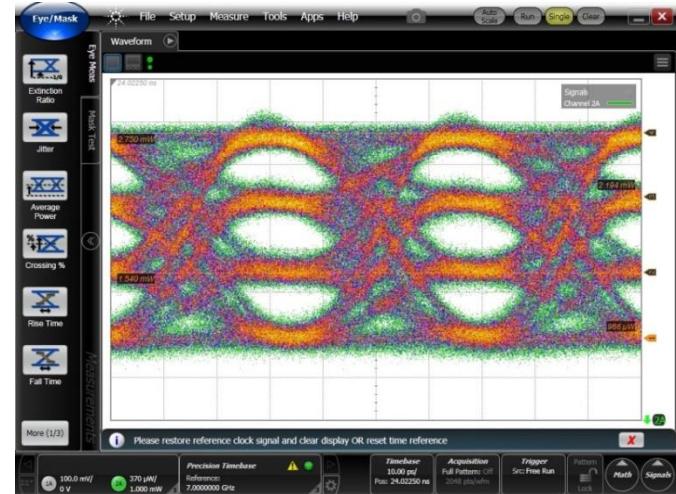


Lane Rate	No. of Lanes	Data Rate	
Gb/s	fiber pairs	$\lambda$	Gb/s
50	1	1	50

# 50GBASE-LR 1310nm SMF Optics



56Gb/s PAM4 optical eye

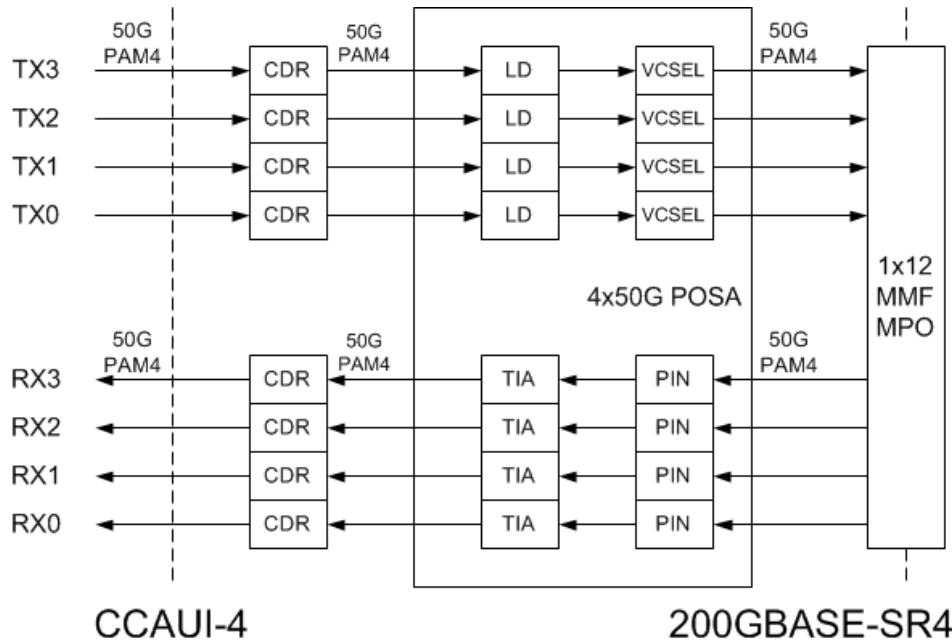


Likely dominant,  
standard form  
factor: SFP28  
w/ 2x LC



Lane Rate	No. of Lanes	Data Rate	
Gb/s	fiber pairs	$\lambda$	Gb/s
50	1	1	50

# 200GBASE-SR4 850nm MMF Optics



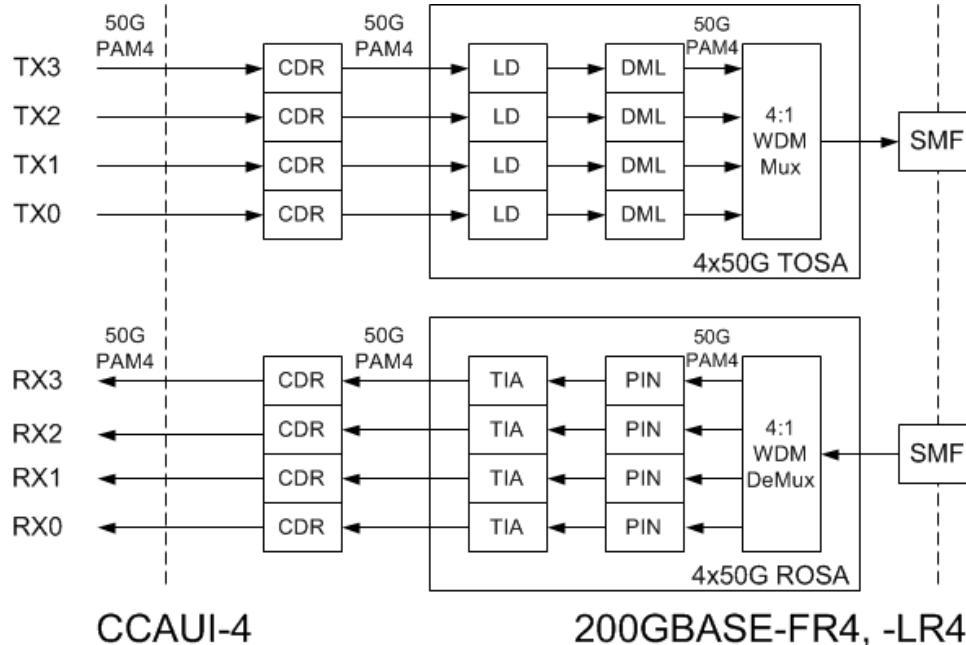
200G PSM4 is parallel SMF version

Likely dominant, standard form factor: QSFP28 w/ MPO



Lane Rate	No. of Lanes	Data Rate	
Gb/s	fiber pairs	$\lambda$	Gb/s
50	4	1	200

# 200GBASE-LR4 1310nm SMF Optics



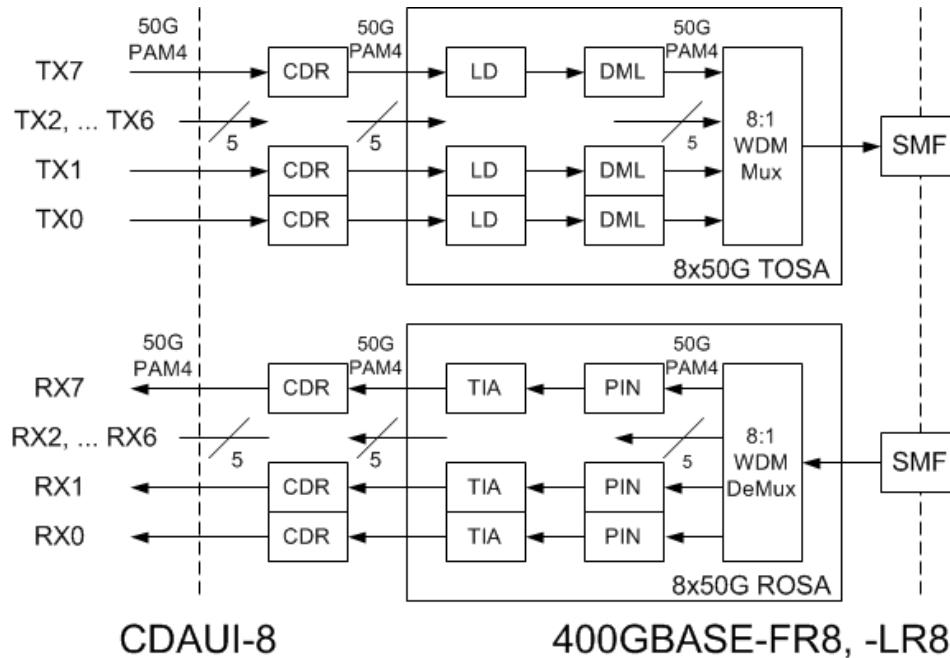
200G SWDM4  
is WDM MMF  
version

Likely dominant,  
standard form  
factor: QSFP28  
w/ 2x LC

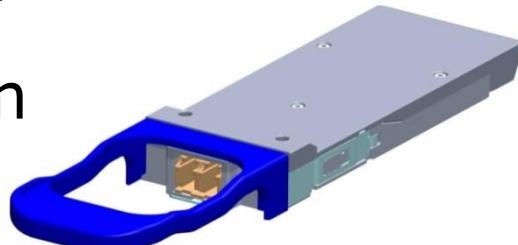


Lane Rate		No. of Lanes		Data Rate	
Gb/s	fiber pairs		$\lambda$	Gb/s	
50	1		4	200	

# 400GBASE-LR8 1310nm SMF Optics



1<sup>st</sup> 400G SMF  
standard form  
factor: CFP8  
w/ 2x LC



Lane Rate	No. of Lanes	Data Rate	
Gb/s	fiber pairs	$\lambda$	Gb/s
50	1	8	400

# 50 Gb/s Lane Optical Interfaces

Lane Rate	No. of Lanes	Data Rate	SW code	LW code	
Gb/s	fiber pairs	$\lambda$	Gb/s	(MMF)	(SMF)
50	1	1	50	SR	LR
50	1	2	100	<i>SWDM2</i>	LR2
50	4	1	200	SR4	<i>PSM4</i>
50	1	4	200	<i>SWDM4</i>	FR4, LR4
50	1	8	400		<b>FR8, LR8</b>

IEEE standards in BOLD; MSA or proprietary in *ITALICS*

# Outline

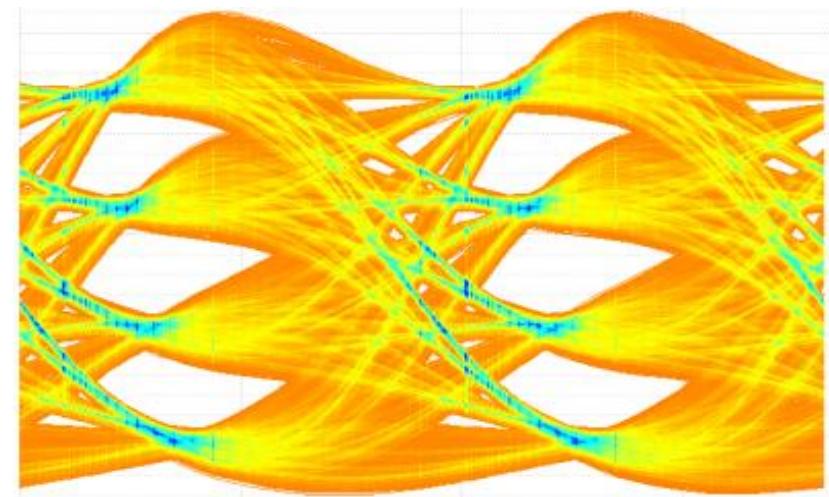
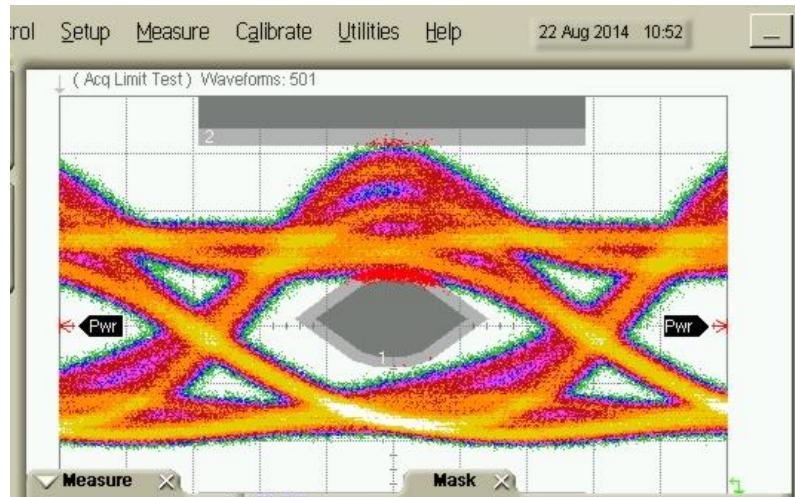
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# 50 Gb/s PAM4 LD Requirements

- DML has different ON vs. OFF damping behavior
- More severe problem for PAM4 than NRZ

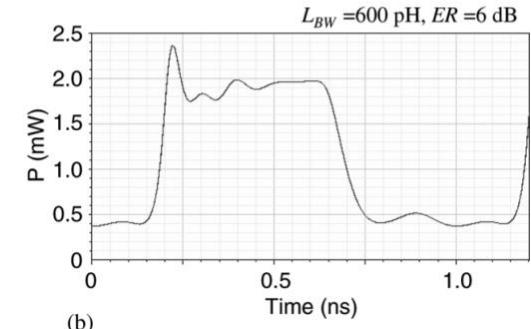
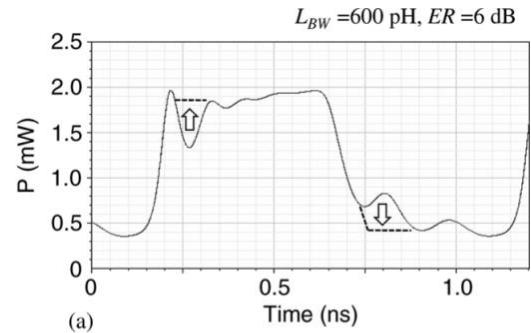
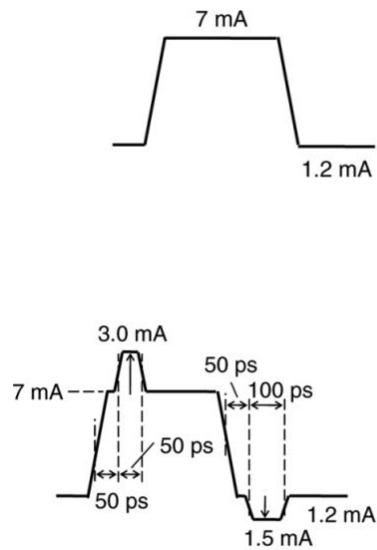
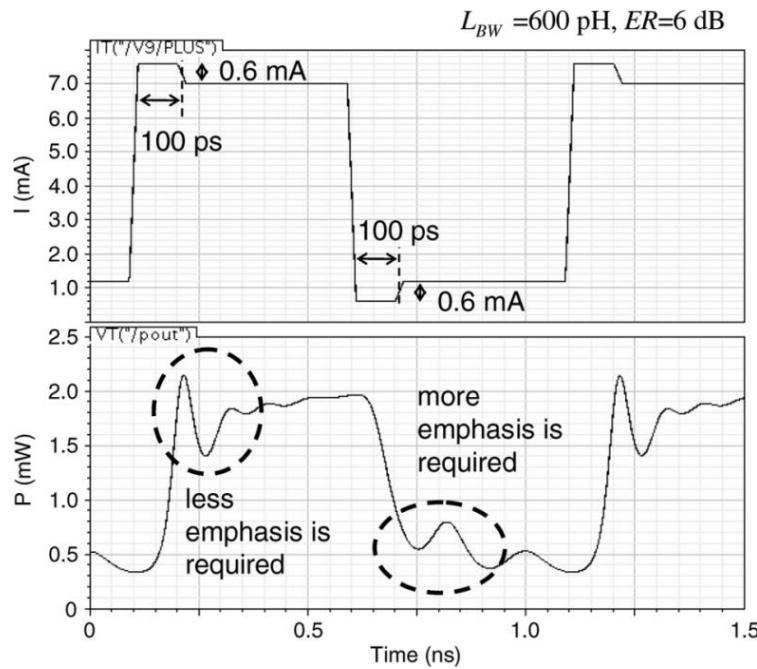
26 Gb/s NRZ optical DCA eye    52 Gb/s PAM4 optical sim. eye



- Requires high-speed LD nonlinear compensation
- Requires linear transfer function LD to support multi-levels at similar low power as NRZ LD

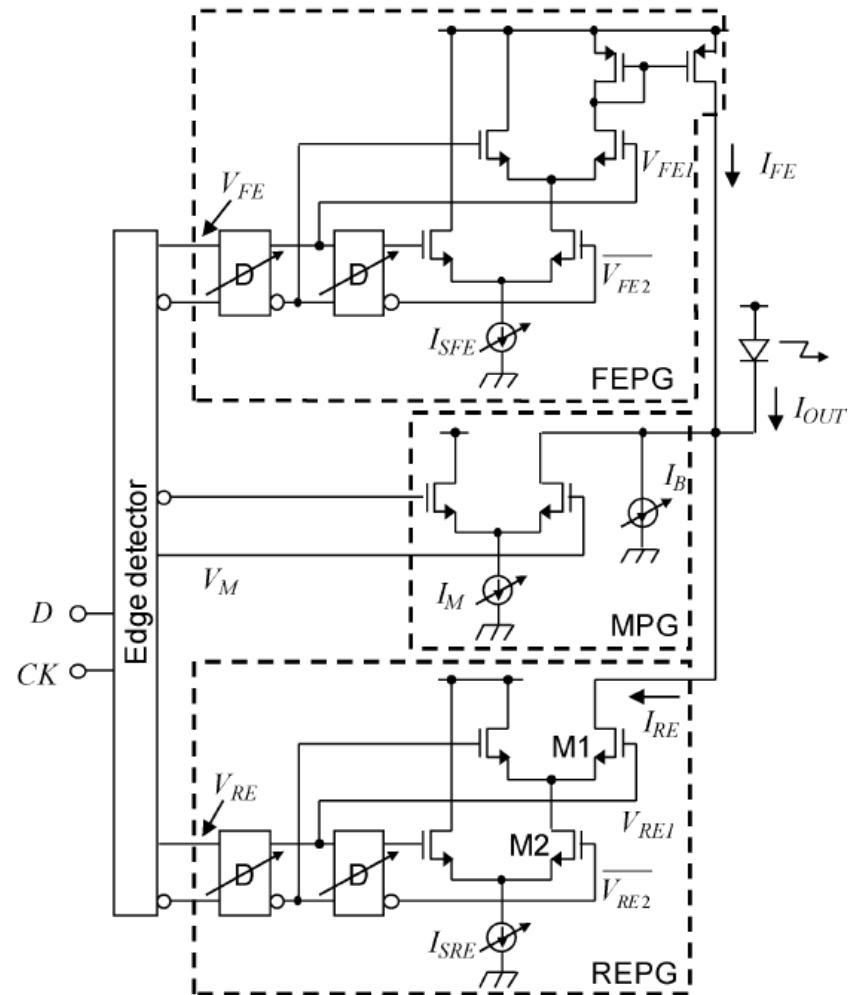
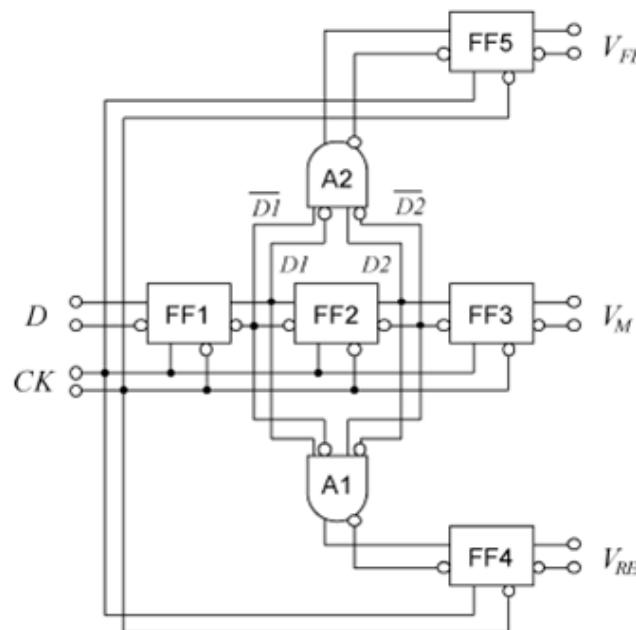
# LD Nonlinear Pre-distortion

## 10 Gb/s LD nonlinear pre-distortion example [5]



# LD Nonlinear Pre-distortion

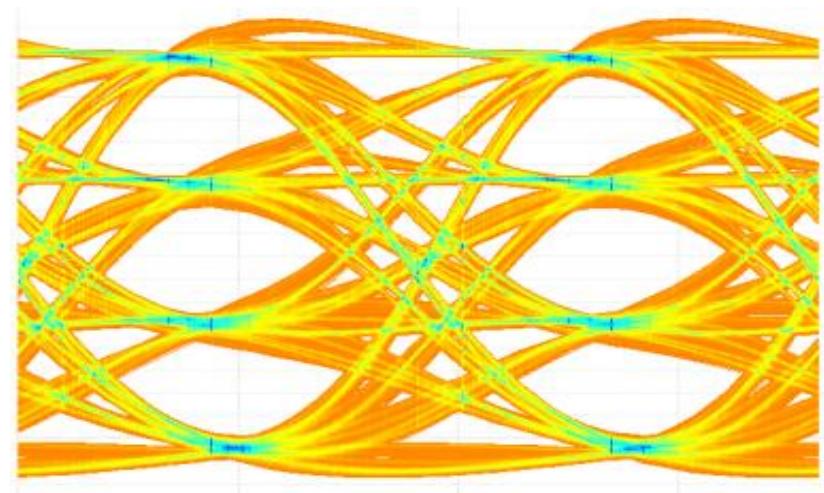
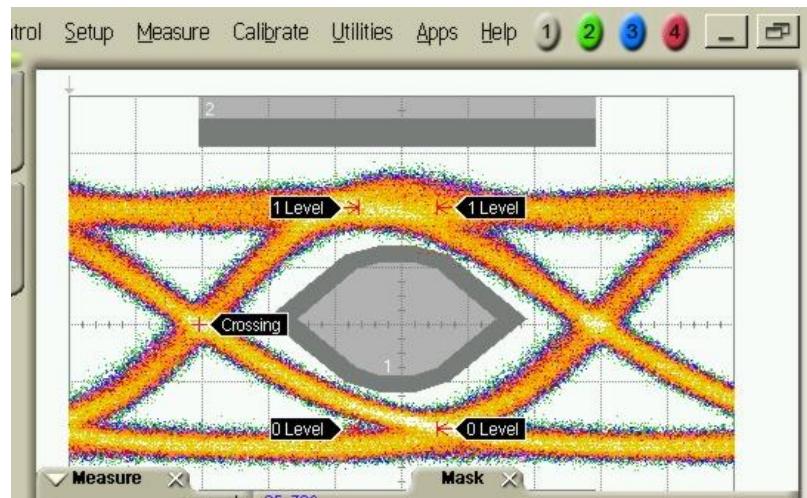
Example edge detector circuit that distinguishes rising and falling edges to select different ON and OFF compensation



# 50 Gb/s PAM4 LD Example

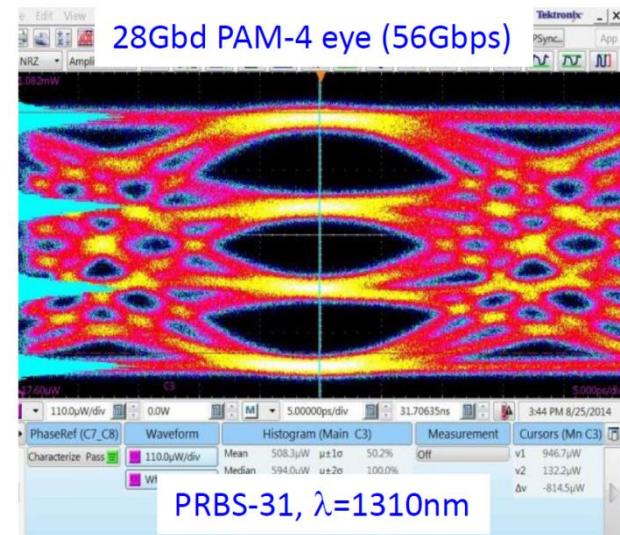
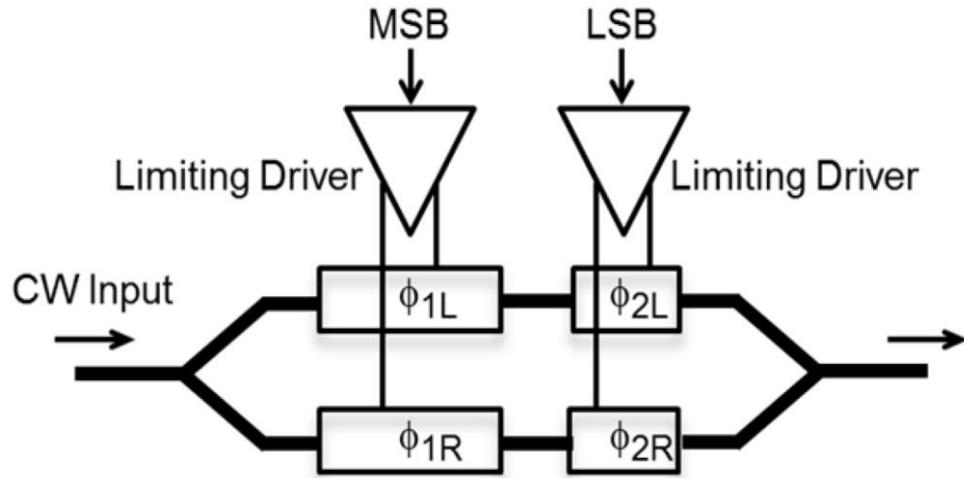
- DML eyes using LD with nonlinear compensation followed by linear transfer function
- Enables use of existing 25G DMLs for PAM4

26 Gb/s NRZ optical DCA eye    52 Gb/s PAM4 optical sim. eye



# 50 Gb/s PAM4 External Modulator

- DML alternative is Continuous Wave (CW) laser w/ linear Si Mach-Zehnder (MZ) modulator [2]
- Cascading binary weighted modulators, driven separately by NRZ bits, creates an optical DAC
- Ex. SiPIC with two NRZ modulator drivers [8]



- Major drawback is low output optical power

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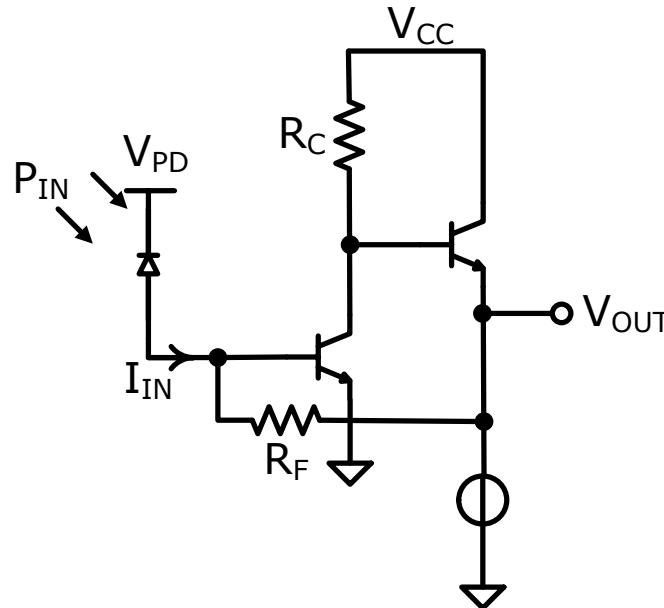
# 50 Gb/s PAM4 TIA Requirements

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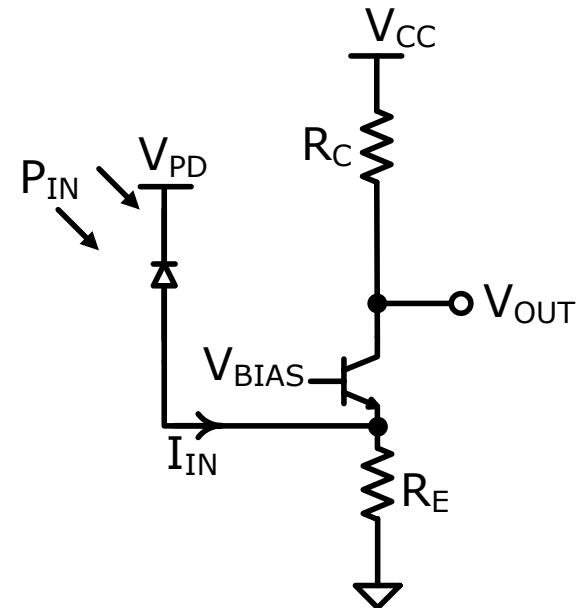
- 50G PAM4 OSNR  $\approx$  25G NRZ OSNR - 5dB
- 50G PAM4 RX<sub>sens</sub>  $\approx$  25G NRZ Rx<sub>sens</sub> - 5dB [4]
- Ex. requirements
  - 100GBASE-LR4 RX<sub>sens</sub> = -10.6dBm OMA  
@1e-12 BER w/o FEC
  - 400GBASE-LR8 RX<sub>sens</sub> = -15.1dBm OMA  
@2e-4 BER w/ KP4 FEC
- PAM4 receiver linearity requirement:
  - THD at Nyquist freq. < 4% over the full dynamic range
  - Requires higher open loop gain, so requires higher open loop bandwidth vs. NRZ

# TIA Topologies

Shunt Feedback (SFB)



Common Base (CB)



TIA parameter	SFB	CB
Trans-impedance	$-R_f$	$R_C$
Input Impedance	$\frac{R_f}{1 + g_m R_C}$	$\frac{1}{g_m}$

# TIA Topologies Sensitivity Comparison

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## □ Shunt Feedback (SFB)

$$\langle i_{in,Total} \rangle^2 = \frac{4kT}{R_f} + 4kTr_b \left( \frac{1}{R_f^2} + \omega^2 C_{PD}^2 \right) + \frac{2qI_C}{g_m^2} \left( \frac{1}{R_f^2} + \omega^2 C_T^2 \right) + \frac{4kT}{R_C} \left( \frac{1}{R_f^2} + \omega^2 C_T^2 \right)$$

## □ Common Base (CB)

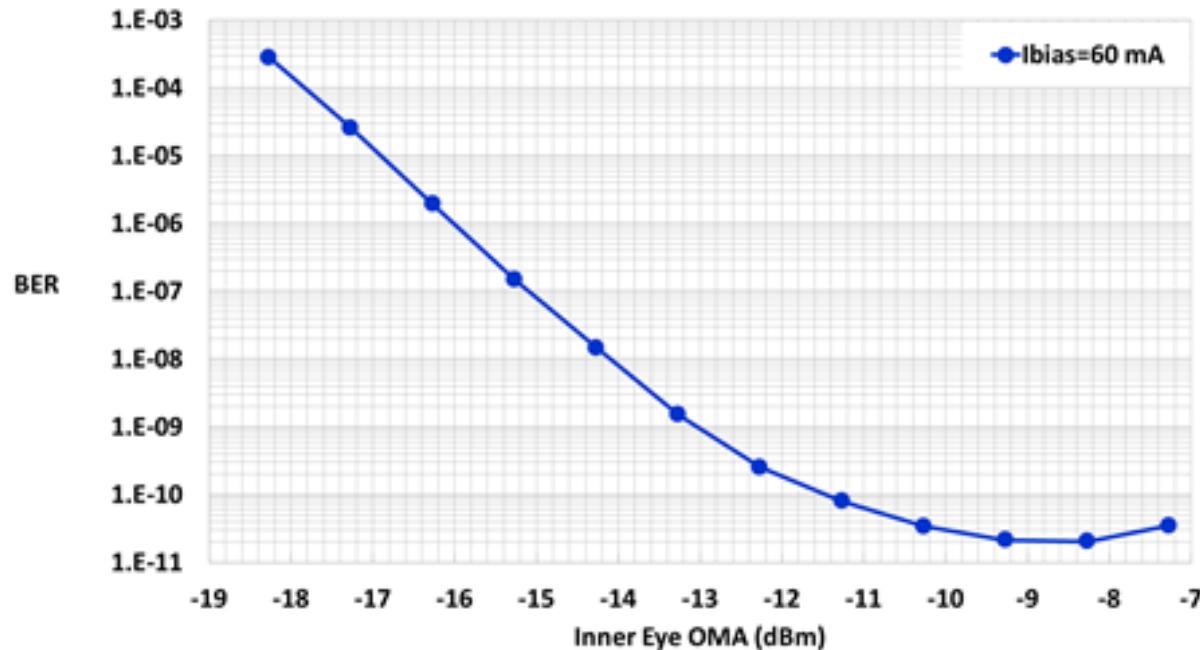
$$\langle i_{in,Total} \rangle^2 = \frac{4kT}{R_E} + 4kTr_b \left( \frac{1}{R_E^2} + \omega^2 C_{PD}^2 \right) + \frac{2qI_C}{g_m^2} \left( \frac{1}{R_E^2} + \omega^2 C_T^2 \right) + \frac{4kT}{R_C} \left( 1 + \frac{\omega^2 C_T^2}{g_m^2} \right)$$

- Given a fixed  $V_{\text{supply}}$ , SFB operates at higher  $I_C$  than CB because it has no  $V$  drop across  $R_E$ 
  - Higher  $I_C$  lowers transistor collector noise  $2qI_C$ , since it is being divided by  $g_m^2$
  - Higher  $I_C$  enables larger transistor area reducing  $r_b$  without sacrificing  $f_T$
- SFB RX<sub>sens</sub> is  $\sim 1.5 \text{dB} > \text{CB}$ , so better for PAM4

# 50 Gb/s PAM4 TIA Example

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- 56 Gb/s DML TX source as shown on page 28
- $\text{RX}_{\text{sens}} \approx -17.5 \text{ dBm OMA} @ 2e-4 \text{ BER (KP4 FEC)}$
- Less margin than for 25 Gb/s NRZ optics
- Error floor makes FEC mandatory



# Outline

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- Wireline Overview
- 10 Gb/s Lane Optical Interfaces
- 25 Gb/s Lane Optical Interfaces
- 50 Gb/s Modulation Selection
- 50 Gb/s Lane Optical Interfaces
- Laser Driver IC
- Trans-Impedance Amplifier IC
- **Clock Data Recovery IC**
- ADC/DSP IC
- Summary
- References

# 50 Gb/s PAM4 CDR Requirements

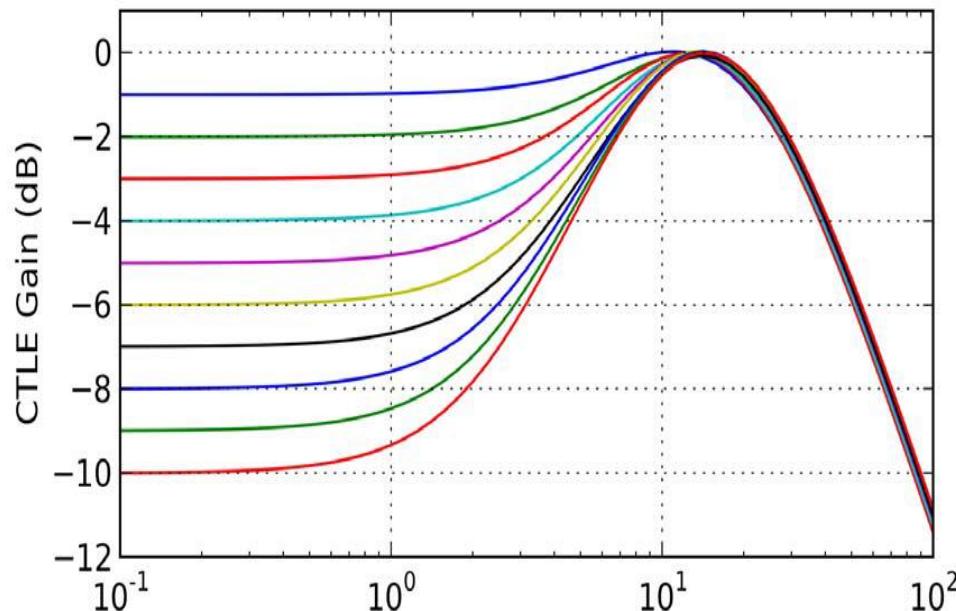
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- 1<sup>st</sup> 50 Gb/s PAM4 lane Ethernet standard is for 400G (8x) and requires adaptive receiver equalization to close optical link budget and eliminate error floors [3]
- For electrical links with channel loss up to ~10dB at Nyquist/2, CTLE is sufficient [13]
- For electrical links with channel loss >10dB, DFE is required, increasing CDR power
- For optical links, 5 to 9-tap T/2-spaced FFE demonstrated to be sufficient [14]
- IEEE 802.3 is discussing the exact normative Eq. for 50G PAM4 optical links; detailed specs to be completed in 2016

# CTLE for Electrical Links

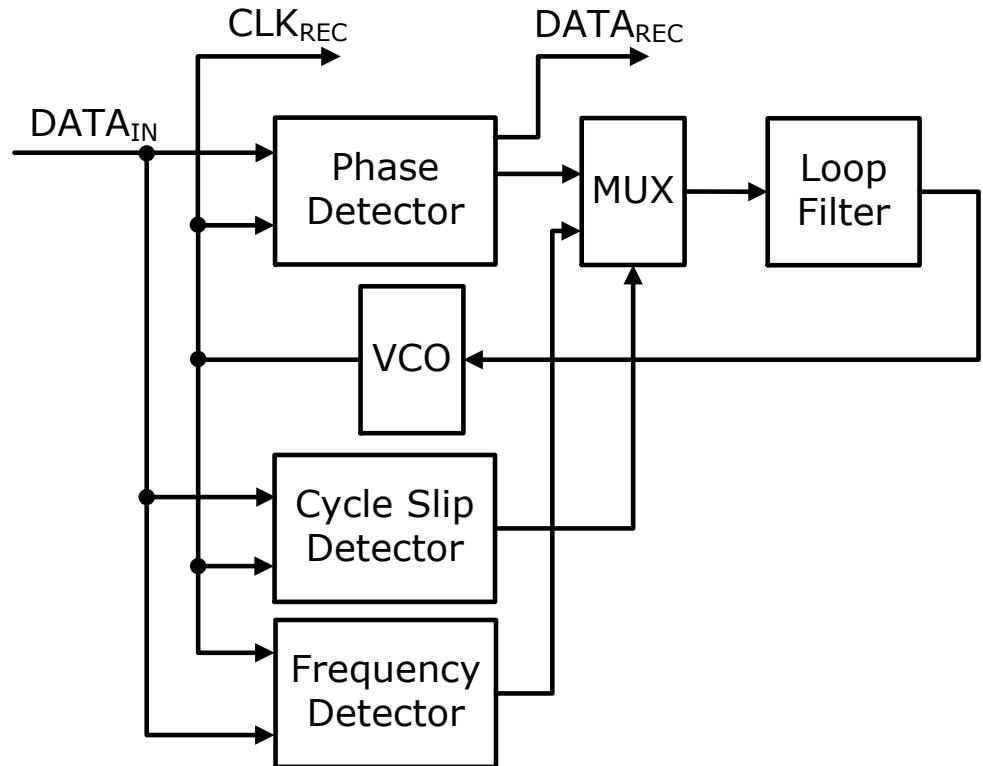
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- Continuous Time Linear Equalizer (CTLE) is used to compensate for channel loss up to 10dB
- Example CTLE characteristics specified by IEEE for 25G NRZ electrical links [8]



# CDR Reference-less Design

- ☐ Lowest power CDR design is reference-less by eliminating external reference clock generation circuits and internal per lane phase rotators
- ☐ Example CDR architecture



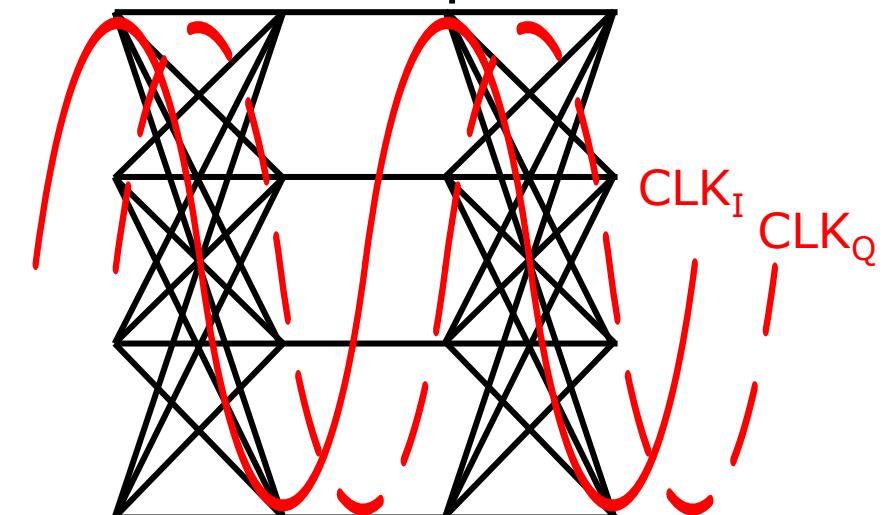
# CDR Phase Detector (PD)

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- Hogge phase detector [9]
  - Linear
  - $T/2$  wide spaced phase correction pulses
  - better jitter performance
  - higher power
- Alexander phase detector [10] [11]
  - non-linear
  - $T$  wide spaced phase correction pulses
  - Digital in nature
  - Lower jitter performance
  - Lower power

# CDR Phase-Frequency Detector (PFD)

- PD has limited pull-in range requiring frequency detector to get within CDR loop bandwidth
- Preferred PFD: Pottbacker [12]
- Drawback is degradation in presence of large amount of DJ since data is used to sample in-phase and quadrature clock in 10ps window
- PAM4 eye has large amount of DJ requiring transition filtering increasing CDR power



# 50 Gb/s PAM4 Analog CDR Example

- 56G Gb/s PAM4 SerDes Transceiver ex. [13]
  - 3 separate decision paths significantly increase PAM4 CDR circuit complexity/area/power
  - Front-end linear Eq. sufficient for optical links
  - DFE Eq. required for higher loss electrical links
-

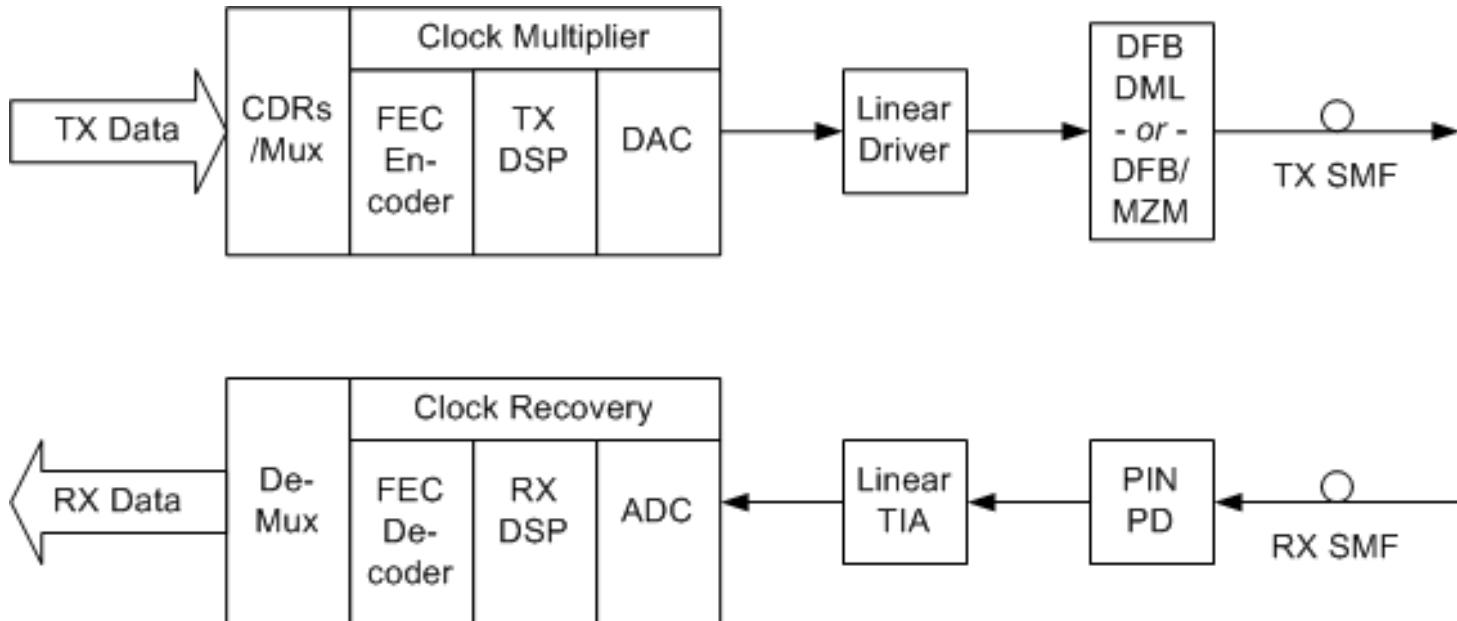
# Outline

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# 50 Gb/s PAM4 ADC/DSP CDR

- Alternative to analog CDR is digital CDR implemented in CMOS DSP with integrated ADC
- Example block diagram shows RX ADC and DSP<sup>1</sup> and optional TX DSP<sup>1</sup> and DAC<sup>1</sup> [1]



<sup>1</sup> Not discussed in this presentation

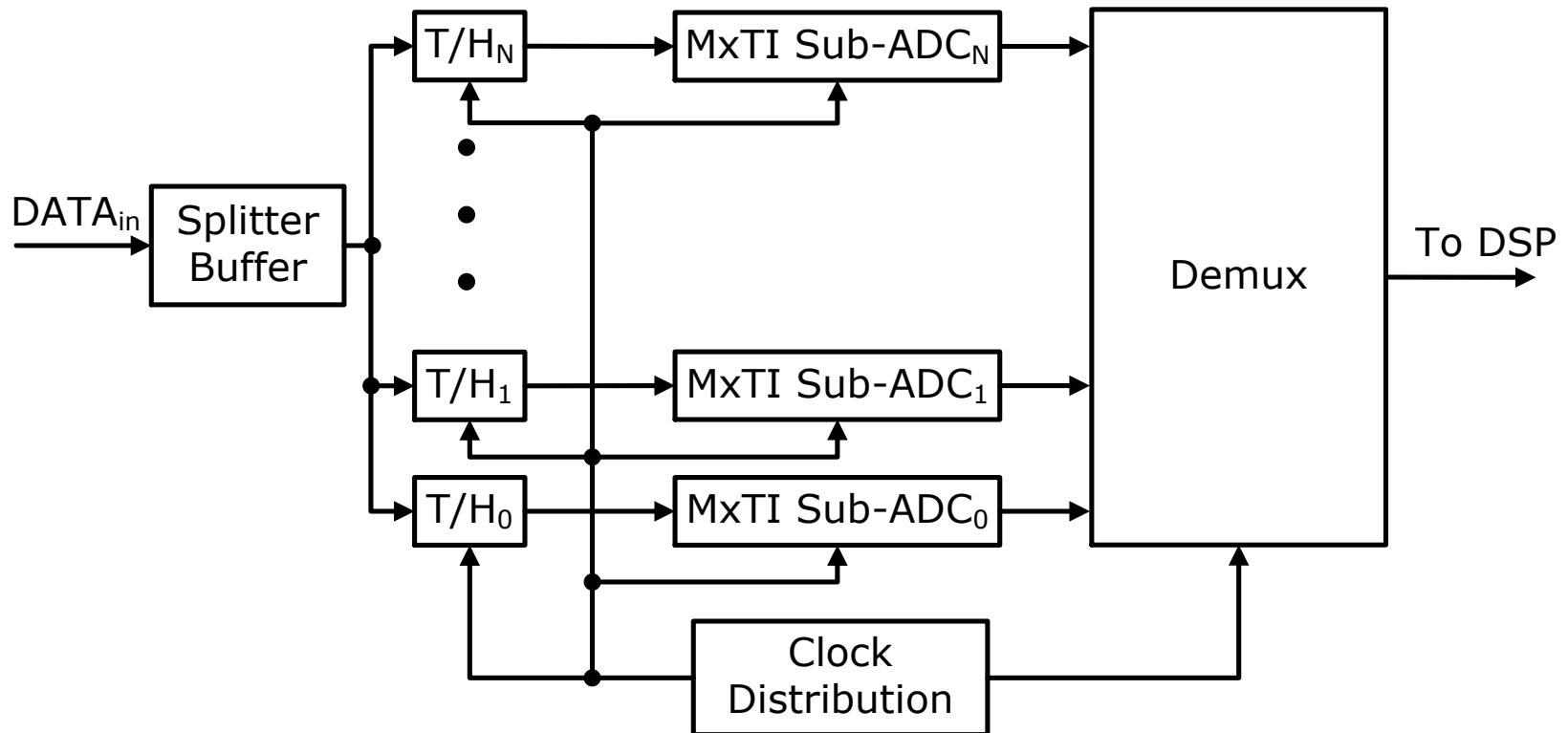
# 50 Gb/s PAM4 ADC/DSP Requirements

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- Continuous-time filter for ADC anti-aliasing
- Baud-rate T-sampling ADC for lowest Eq. power
  - Requires robust timing recovery for low penalty
- High loss links require pre-equalization to enable DSP clock recovery
  - Increases latency
  - Requires low loop bandwidth
  - Reduces jitter tolerance
- DSP supports higher number of FFE and/or DFE taps beyond required for CDR to close the link, enabling minimization of optical link penalties

# 50 Gb/s ADC Implementation

- Time-interleaved SAR is ideal ASIC block [16]
- N master T/H (Track/Hold) x M sub-ADC T/H time-interleaved SAR ADC example:



# 50 Gb/s SAR ADC

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- Splitter buffer determines bandwidth and THD
- Clock timing to N master T/H's and Master T/H gain-error determines resolution
- As CMOS scales, process variation and mismatch limit ADC performance improvements
- 8-bit nominal SAR ADC example [17]
  - 40nm CMOS
    - 6-bit ENOB w/ 16GHz BW
    - 63 GS/s @ 1.25W
  - 28nm CMOS
    - 6-bit ENOB w/ 25GHz BW
    - 56GS/s @ 0.8W
    - 28GS/s @ 0.4W

# Summary

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- 50 Gb/s PAM4 is the next high-volume short-reach interconnect lane technology
- 50 Gb/s (1x), 100 Gb/s (2x), 200 Gb/s (4x) and 400 Gb/s (8x) data rates will be supported
- 50 Gb/s PAM4 lanes requires adaptive Eq. (analog or digital) and FEC
- Circuit design challenges and opportunities
  - Linear laser driver
  - Linear trans-impedance amplifier
  - Si modulator driver and modulator
  - Multi-level adaptive clock-data recovery
  - High-speed ADC

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